

Time harmonic fields

(plane, spherical and cylindrical waves; solution of wave equation or radiation condition; wave velocity; Poynting theorem)

Hereon we will concentrate on time harmonic fields (i.e., sinusoidal) of the form

$$\cos(\omega t - kx), \sin(\omega t - ky), \cos(\omega t - kr) \text{ or } \sin(\omega t - kp) \text{ (cylindrical), etc.}$$

If the field is not in this form, we can invoke Fourier Series to rewrite the given field as

$$E(t, x=0) = \sum (A_n \cos n\omega_0 t + B_n \sin n\omega_0 t) = \sum c_n e^{jn\omega t}$$

where ω_0 = the fundamental frequency. Thus, any time-domain field can be represented as a superposition of *harmonics*, and for our analysis we can consider each harmonic component separately. That is, we can instead deal with phasor quantities as done in circuits.

In general, we can express \mathbf{E} as

$$\mathbf{E}(x, y, z; t) = \text{Re (or Im)} \{ \mathbf{E}(x, y, z) e^{j\omega t} \}$$

Note that we are adopting the $e^{+j\omega t}$ convention (although in many physics books the $e^{-j\omega t}$ convention is used). This representation should be compared to

$$V(t) = \text{Re} \{ V_{\text{phasor}} e^{j\omega t} \}$$

Here, the phasor will be the field components, viz.

$$E_{\frac{y}{z}}(x, y, z, t) = \text{Re (or Im)} \left\{ E_{\frac{y}{z}}(x, y, z) e^{j\omega t} \right\}$$

with

$$E_{\frac{y}{z}}(x, y, z) = \left| E_{\frac{y}{z}} \right| e^{j\phi_{\frac{y}{z}}} = E_{x0} e^{j\phi_{\frac{y}{z}}} e^{j\phi_{y0}}$$

Thus, the general form (hereon $\text{Re}\{ \}$ and $e^{j\omega t}$ conventions will be adopted) will be

$$\begin{aligned} \mathbf{E}(x, y, z, t) \text{ or } \mathfrak{E}(x, y, z, t) &= \text{Re} \{ \mathbf{E}(x, y, z) e^{j\omega t} \} \\ &= \hat{x}E_{x0} \cos(\omega t + \phi_x) + \hat{y}E_{y0} \cos(\omega t + \phi_y) + \hat{z}E_{z0} \cos(\omega t + \phi_z) \end{aligned}$$

(see p. 15 of Harrington). The coefficients E_{x0} , E_{y0} , E_{z0} represent peak values of the harmonic field. Harrington instead considers E_{x0} , E_{y0} , E_{z0} to be RMS values. This difference will simply impact the formulas for the time-average power/energy.

Example: (see p. 22 of Harrington)

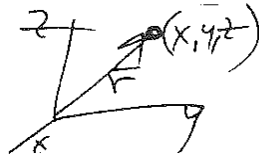
$$\frac{1}{2} \iiint_V \sigma |\mathbf{E}|^2 dv \quad \text{average power dissipated in } V \text{ (no } \frac{1}{2} \text{ in Harrington)}$$

$$\frac{1}{4} \iiint_V \epsilon_0 \epsilon_r |\mathbf{E}|^2 dv \quad \text{average electric energy stored in } V \text{ (}\frac{1}{2} \text{ instead of } \frac{1}{4} \text{ in Harrington)}$$

We can readily rewrite Maxwell's equations in phasor form by noting that

$$\frac{\partial \mathbf{E}}{\partial t} = j\omega \mathbf{E}(x, y, z, t) = j\omega \mathbf{E}(\mathbf{r})$$

where $\mathbf{r} = x\hat{x} + y\hat{y} + z\hat{z}$ is the spherical or position vector.



Basically, we set

$$\frac{\partial}{\partial t} \longrightarrow j\omega$$

when dealing with sinusoidal/phasor quantities.

$\nabla \times \mathbf{E} = -j\omega \mathbf{H}$	$\nabla \times \mathbf{E} = -\mathbf{M} - j\omega \mu \mathbf{H} \quad (M \rightarrow \text{V/m}^2)$
$\nabla \times \mathbf{H} = \mathbf{J} - j\omega \epsilon \mathbf{E}$	$\nabla \times \mathbf{H} = \mathbf{J} + j\omega \epsilon \mathbf{E} \quad (J \rightarrow \text{A/m}^2)$
$\nabla \cdot \mathbf{J} = -j\omega \rho$	$\nabla \cdot \mathbf{J} = -j\omega \rho$
	$\nabla \cdot \mathbf{M} = -j\omega \rho_m \quad (\rho_m = \text{magn. charges})$
$\nabla \cdot (\mu \mathbf{H}) = 0$	$\nabla \cdot \mu \mathbf{H} = \rho_m \quad (\rho_m \rightarrow \text{Wb/m}^3)$
$\nabla \cdot (\epsilon \mathbf{E}) = \rho$	$\nabla \cdot \epsilon \mathbf{E} = \rho \quad (\rho \rightarrow \text{C/m}^3)$
$\mathbf{D} = \epsilon \mathbf{E}$	$\mathbf{D} = \epsilon \mathbf{E} \quad (D \rightarrow \text{C/m}^2)$
$\mathbf{B} = \mu \mathbf{H}$	$\mathbf{B} = \mu \mathbf{H} \quad (B \rightarrow \text{Wb/m}^2)$

Magnetic Current

For mathematical convenience we may also assume the presence of magnetic currents as added sources. That is, we accept the possibility that the south and north poles of a magnet can exist in isolation. If magnetic charges exist (caused by electron spins) then

$$\iint \mathbf{B} \cdot d\mathbf{s} = \iiint_V \rho_m dv \Leftrightarrow \nabla \cdot \mathbf{B} = \rho_m \quad \text{or} \quad \nabla \cdot \mu \mathbf{H} = \rho_m$$

The corresponding continuity equation for magnetic charges is

$$\nabla \cdot \mathbf{M} = -j\omega \rho_m$$

and the modified Maxwell's equation is

$$\nabla \times \mathbf{E} = -\mathbf{M} - j\omega \mu \mathbf{H}$$

such that

$$\nabla \cdot (\nabla \times \mathbf{E}) = 0 = -\nabla \cdot \mathbf{M} - j\omega \nabla \cdot \mu \mathbf{H}$$

satisfies the continuity relation. In practice, magnetic currents are useful for modeling and representing jumps in the electric field.

Complex permittivity

- Goal is to incorporate σ into ϵ

Taking a look at Ampère's Law, we have

$$\nabla \times \mathbf{H} = \mathbf{J}_i + \underbrace{\frac{\sigma}{j\omega} \mathbf{E}}_{\mathbf{J}_c} + j\omega \epsilon \mathbf{E}$$

We see that we can rewrite this as

$$\begin{aligned}\nabla \times \mathbf{H} &= \mathbf{J}_i + (\sigma + j\omega \epsilon) \mathbf{E} \\ &\downarrow \\ &j\omega \left(\frac{\sigma}{j\omega} + \epsilon \right) \mathbf{E} \\ &\downarrow \\ &\text{new } \epsilon \rightarrow \dot{\epsilon} \\ \nabla \times \mathbf{H} &= \mathbf{J}_i + j\omega \dot{\epsilon} \mathbf{E} \\ \dot{\epsilon} &= \epsilon + \frac{\sigma}{j\omega} = \epsilon' - j\epsilon'' = \epsilon' (1 - j \tan \delta_e) \\ \tan \delta_e &= \text{loss tangent} = \frac{\epsilon''}{\epsilon'} \\ \tan \delta_e &= 0 \quad \leftarrow \text{lossless media (non-conducting)} \\ \tan \delta_e &\neq 0 \quad \leftarrow \text{lossy media } (\sigma \neq 0)\end{aligned}$$

Hereon, we will simply write Ampère's Law as

$$\nabla \times \mathbf{H} = \mathbf{J}_i + j\omega \epsilon \mathbf{E}$$

and drop the “.” over ϵ . Thus, whenever we write ϵ in a time harmonic/phasor equation, we will imply that it has real and imaginary parts

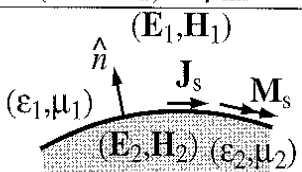
$$\begin{aligned}\epsilon &= \epsilon' - j\epsilon'' && (\epsilon', \epsilon'') \text{ are given} \\ &= \epsilon' (1 - j \tan \delta_e) && (\epsilon', \delta_e) \text{ are given} \\ &= \epsilon_{\text{real}} + \frac{\sigma}{j\omega} && (\epsilon_{\text{real}}, \sigma) \text{ given}\end{aligned}$$

(Any of these combinations can be given.) Similarly,

$$\begin{aligned}\epsilon_r &= \frac{\epsilon}{\epsilon_0} = \epsilon'_r - j\epsilon''_r && (\text{complex relative permittivity}) \\ &= (\epsilon_r)_{\text{real}} + \frac{\sigma}{j\omega \epsilon_0}\end{aligned}$$

Likewise for magnetic material

$$\begin{aligned}\mu &= \mu' - j\mu'' \\ &= \mu' (1 - j \tan \delta_m) && \delta_m : \text{magnetic loss tangent} \\ &= \mu_{\text{real}} + \frac{\sigma_m}{j\omega}\end{aligned}$$

Final Equations for Time Harmonic Fields	Boundary Conditions
$\nabla \times \mathbf{E} = -\mathbf{M}_i - j\omega\mu\mathbf{H}$ $\nabla \times \mathbf{H} = \mathbf{J}_i + j\omega\epsilon\mathbf{E}$ $\nabla \cdot \mathbf{J}_i = -j\omega\rho$ $\nabla \cdot \mathbf{M}_i = -j\omega\rho_m$	$\hat{n} \times (\mathbf{E}_1 - \mathbf{E}_2) = -\mathbf{M}_s$ (impressed $\hat{n} \times (\mathbf{H}_1 - \mathbf{H}_2) = +\mathbf{J}_s$ surface charges)
$\nabla \cdot \epsilon\mathbf{E} = \rho$ $\nabla \cdot \mu\mathbf{H} = \rho_m$	$\hat{n} \cdot (\mathbf{D}_1 - \mathbf{D}_2) = \rho_s$ $\hat{n} \cdot (\mathbf{B}_1 - \mathbf{B}_2) = \rho_{ms}$
$\mathbf{B} = \mu\mathbf{H}$ $\mathbf{D} = \epsilon\mathbf{E}$	
$\epsilon = \epsilon' - j\epsilon'' = \epsilon'(1 - j\tan\delta_e)$ $\quad \quad \quad = \epsilon_{\text{real}} + \sigma/j\omega$ $\mu = \mu' - j\mu'' = \mu'(1 - j\tan\delta_m)$ $\quad \quad \quad = \mu_{\text{real}} + \sigma_m/j\omega$ $\delta_e = \text{electric loss tangent}$ $\delta_m = \text{magnetic loss tangent}$	Special Cases 1. Medium #2 is perfect (electric) conductor $\hat{n} \times \mathbf{H}_1 = \mathbf{J}_s$ $\hat{n} \times \mathbf{E}_1 = 0$ $\hat{n} \cdot \mathbf{B} = 0$ $\hat{n} \cdot \mathbf{D} = \rho_{sy}$ 2. Medium #2 is perfect magnetic conductor (PMC) $\hat{n} \times \mathbf{H}_1 = 0$ $\hat{n} \times \mathbf{E}_1 = -\mathbf{M}_s$ $\hat{n} \cdot \mathbf{B}_1 = \rho_{ms}$ $\hat{n} \cdot \mathbf{D} = 0$

Wave Equations

$$Z_0 = \sqrt{\mu_0/\epsilon}, \quad k_0 = \omega\sqrt{\mu_0\epsilon_0}$$

$$\nabla \times \left(\frac{\nabla \times \mathbf{E}}{\mu_r} \right) - k_0^2 \epsilon_r \mathbf{E} = -jk_0 Z_0 \mathbf{J}_i - \nabla \times \left(\frac{\mathbf{M}_i}{\mu_r} \right)$$

or

$$\frac{1}{\mu_r} \nabla \times \nabla \times \mathbf{E} - k_0^2 \epsilon_r \mathbf{E} + \nabla \left(\frac{1}{\mu_r} \right) \times \nabla \times \mathbf{E} = -jk_0 Z_0 \mathbf{J}_i - \nabla \times \left(\frac{\mathbf{M}_i}{\mu_r} \right)$$

For homogeneous media: $\nabla^2 \mathbf{E} - \nabla(\nabla \cdot \mathbf{E}) + k_0^2 \epsilon_r \mu_r \mathbf{E} = jk_0 Z_0 \mu_r \mathbf{J}_i + \nabla \times \mathbf{M}_i$

Homogeneous and source-free: $\nabla^2 \mathbf{E} + k_0^2 \epsilon_r \mu_r \mathbf{E} = jk_0 Z_0 \mu_r \mathbf{J}_i + \nabla \times \mathbf{M}_i$